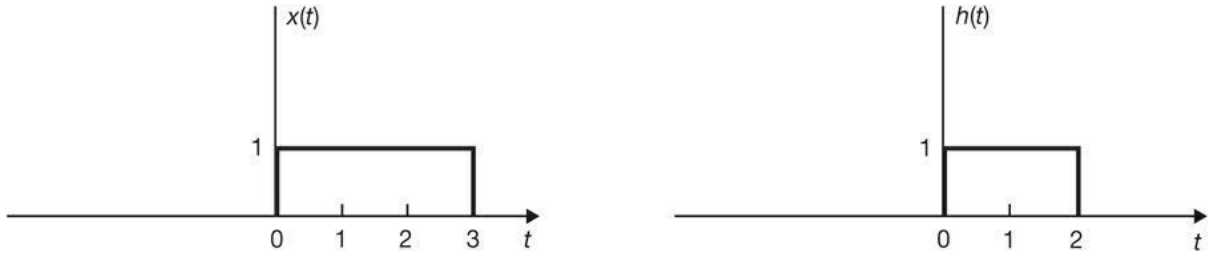


1-03-d tutorial solution

Question 1

2.6 Evaluate $y(t) = x(t) * h(t)$, where $x(t)$ and $h(t)$ are shown in Fig. 2-6, (a) by an analytical technique, and (b) by a graphical method.



SOLUTION

(a)

We first express $x(t)$ and $h(t)$ in functional form: $x(t) = u(t) - u(t - 3)$ $h(t) = u(t) - u(t - 2)$

$$\begin{aligned}
 y(t) &= x(t) * h(t) = \int_{-\infty}^{\infty} x(\tau)h(t - \tau) d\tau \\
 &= \int_{-\infty}^{\infty} [u(\tau) - u(\tau - 3)][u(t - \tau) - u(t - \tau - 2)] d\tau \\
 &= \int_{-\infty}^{\infty} u(\tau)u(t - \tau) d\tau - \int_{-\infty}^{\infty} u(\tau)u(t - 2 - \tau) d\tau \\
 &\quad - \int_{-\infty}^{\infty} u(\tau - 3)u(t - \tau) d\tau + \int_{-\infty}^{\infty} u(\tau - 3)u(t - 2 - \tau) d\tau
 \end{aligned}$$

$$\begin{aligned}
 u(\tau)u(t - \tau) &= \begin{cases} 1 & 0 < \tau < t, t > 0 \\ 0 & \text{otherwise} \end{cases} \\
 u(\tau)u(t - 2 - \tau) &= \begin{cases} 1 & 0 < \tau < t - 2, t > 2 \\ 0 & \text{otherwise} \end{cases} \\
 u(\tau - 3)u(t - \tau) &= \begin{cases} 1 & 3 < \tau < t, t > 3 \\ 0 & \text{otherwise} \end{cases} \\
 u(\tau - 3)u(t - 2 - \tau) &= \begin{cases} 1 & 3 < \tau < t - 2, t > 5 \\ 0 & \text{otherwise} \end{cases}
 \end{aligned}$$

We can express $y(t)$ as:

$$\begin{aligned}
 y(t) &= \left(\int_0^t d\tau \right) u(t) - \left(\int_0^{t-2} d\tau \right) u(t - 2) \\
 &\quad - \left(\int_3^t d\tau \right) u(t - 3) + \left(\int_3^{t-2} d\tau \right) u(t - 5) \\
 &= tu(t) - (t - 2)u(t - 2) - (t - 3)u(t - 3) + (t - 5)u(t - 5)
 \end{aligned}$$

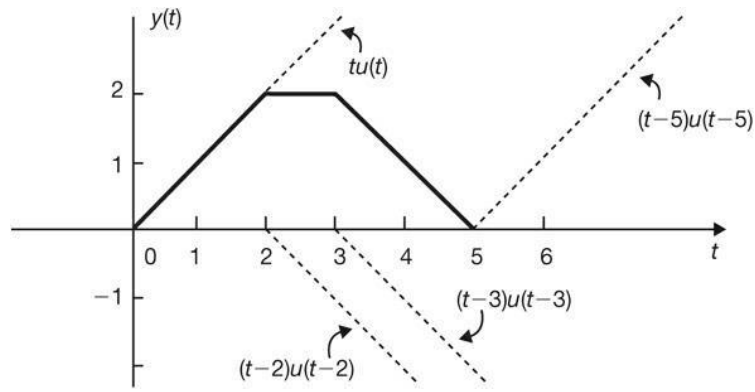


Fig 2-7

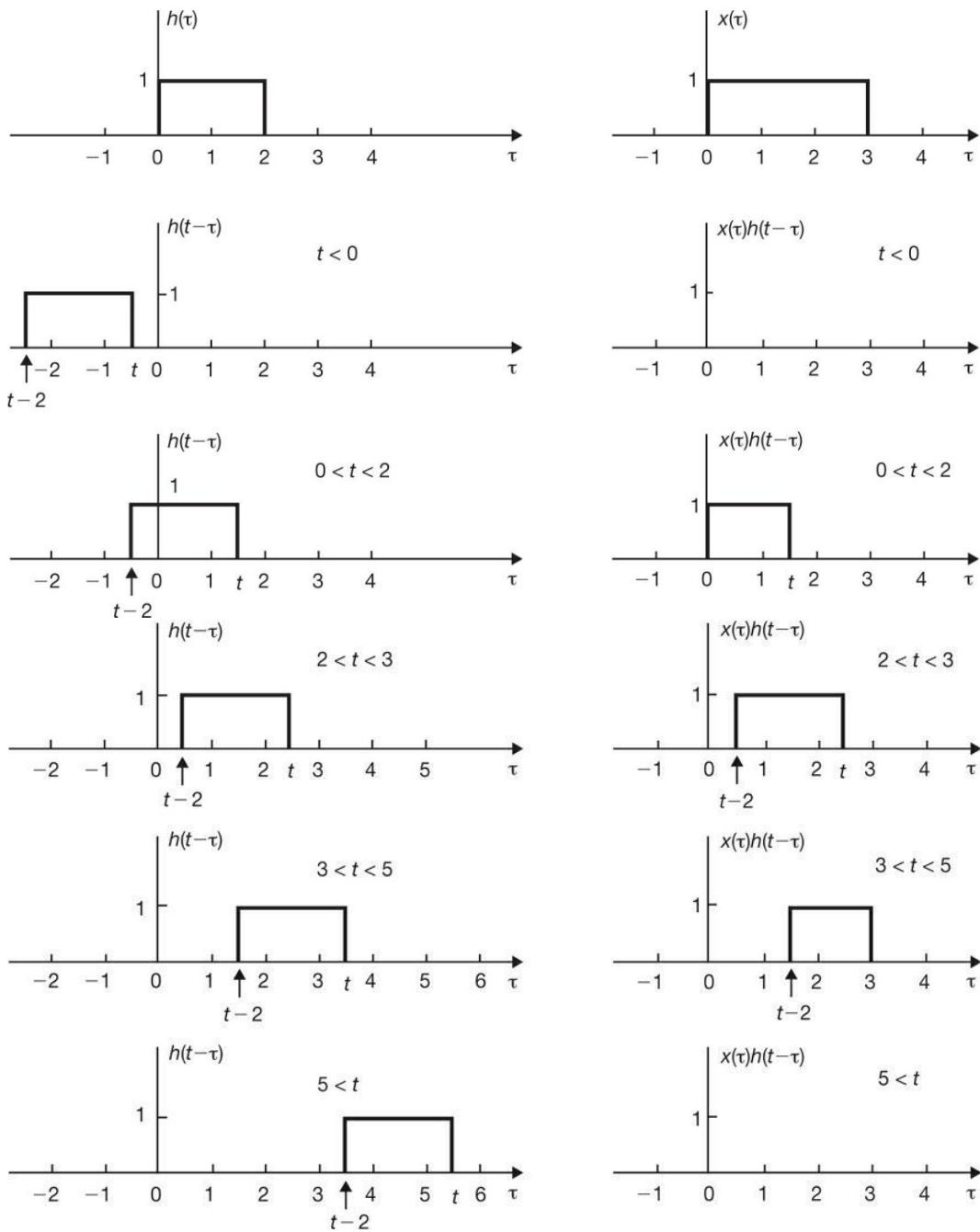


Fig 2-8

(b)

Functions $h(\tau)$, $x(\tau)$ and $h(t - \tau)$, $x(\tau)h(t - \tau)$ for different values of t are sketched in Fig. 2-8. From Fig. 2-8 we see that $x(\tau)$ and $h(t - \tau)$ do not overlap for $t < 0$ and $t > 5$, and hence, $y(t) = 0$ for $t < 0$ and $t > 5$.

For the other intervals, $x(\tau)$ and $h(t - \tau)$ overlap. Thus, computing the area under the rectangular pulses for these intervals, we obtain:

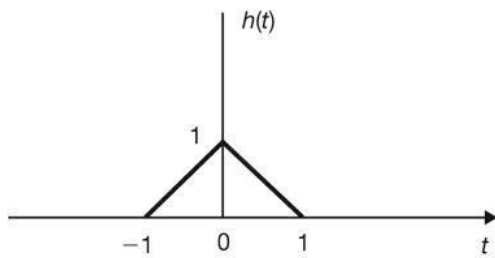
$$y(t) = \begin{cases} 0 & t < 0 \\ t & 0 < t \leq 2 \\ 2 & 2 < t \leq 3 \\ 5 - t & 3 < t \leq 5 \\ 0 & 5 < t \end{cases}$$

Question 2

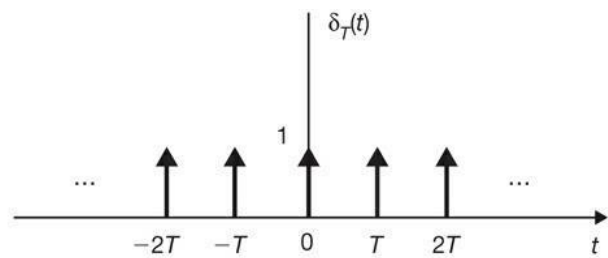
Let $h(t)$ be the triangular pulse shown in Fig. 2-10(a) and let $x(t)$ be the unit impulse train [Fig. 2-10(b)] expressed as

$$x(t) = \delta_T(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT) \quad (2.68)$$

Determine and sketch $y(t) = h(t) * x(t)$ for the following values of T : (a) $T = 3$, (b) $T = 2$, (c) $T = 1.5$.



(a)



(b)

SOLUTION

According to the property of the sampling signal:

$$\begin{aligned} y(t) &= h(t) * \delta_T(t) = h(t) * \left[\sum_{n=-\infty}^{\infty} \delta(t - nT) \right] \\ &= \sum_{n=-\infty}^{\infty} h(t) * \delta(t - nT) = \sum_{n=-\infty}^{\infty} h(t - nT) \end{aligned} \quad (2.69)$$

which is sketched in Fig. 2-11(c). Note that when $T < 2$, the triangular pulses are no longer separated and they overlap.

(a) For $T = 3$, Eq. (2.69) becomes

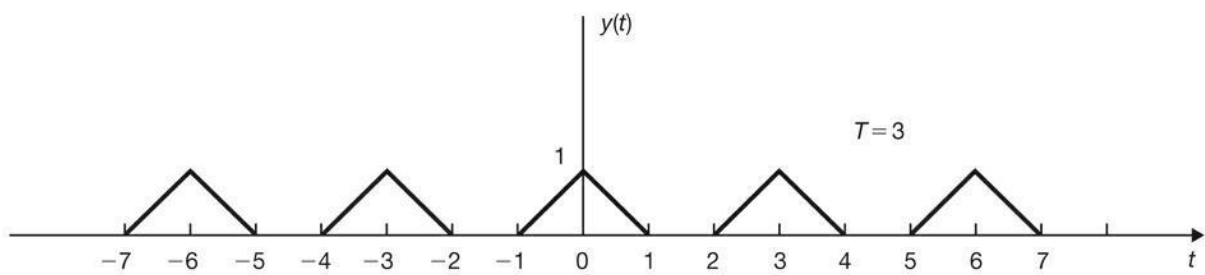
$$y(t) = \sum_{n=-\infty}^{\infty} h(t - 3n)$$

(b) For $T = 2$, Eq. (2.69) becomes

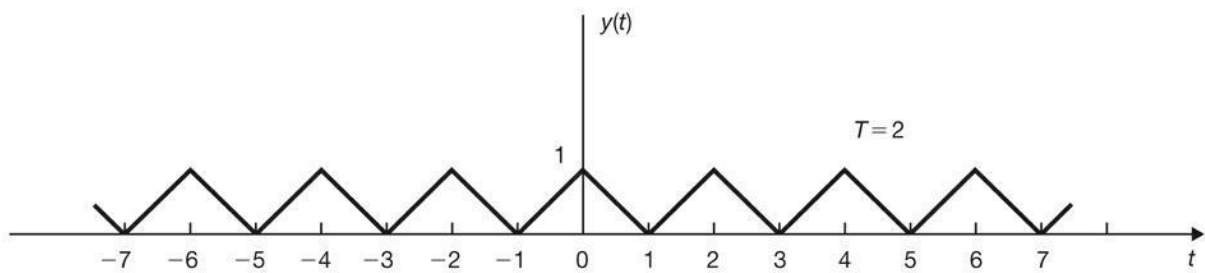
$$y(t) = \sum_{n=-\infty}^{\infty} h(t - 2n)$$

(c) For $T = 1.5$, Eq. (2.69) becomes

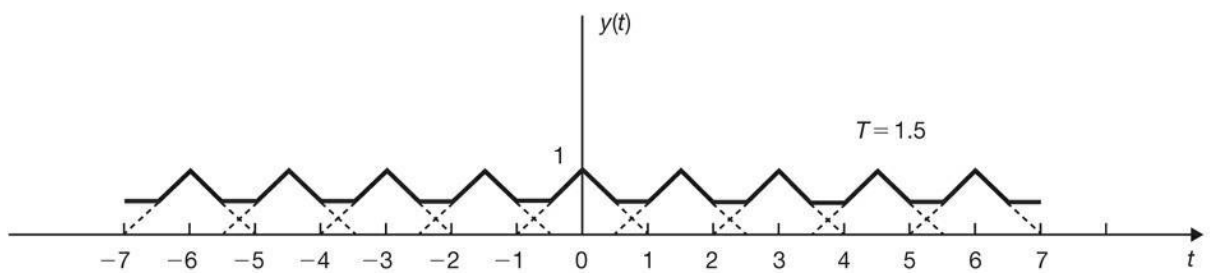
$$y(t) = \sum_{n=-\infty}^{\infty} h(t - 1.5n)$$



(a)



(b)



(c)

Fig 2-11

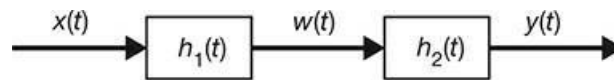
Question 3

2.14. The system shown in Fig. 2-17(a) is formed by connecting two systems in cascade. The impulse responses of the systems are given by $h_1(t)$ and $h_2(t)$, respectively, and

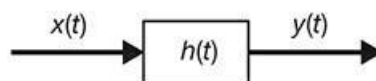
$$h_1(t) = e^{-2t} u(t) \quad h_2(t) = 2e^{-t} u(t)$$

(a) Find the impulse response $h(t)$ of the overall system shown in Fig. 2-17(b).

(b) Determine if the overall system is BIBO stable.



(a)



(b)

SOLUTION

(a) Let $w(t)$ be the output of the first system. By Eq. (2.6)

$$w(t) = x(t) * h_1(t) \tag{2.78}$$

Then we have

$$y(t) = w(t) * h_2(t) = [x(t) * h_1(t)] * h_2(t) \tag{2.79}$$

But by the associativity property of convolution (2.8), Eq. (2.79) can be rewritten as

$$y(t) = x(t) * [h_1(t) * h_2(t)] = x(t) * h(t) \tag{2.80}$$

Therefore, the impulse response of the overall system is given by

$$h(t) = h_1(t) * h_2(t) \tag{2.81}$$

Thus, with the given $h_1(t)$ and $h_2(t)$, we have

$$\begin{aligned} h(t) &= \int_{-\infty}^{\infty} h_1(\tau) h_2(t - \tau) d\tau = \int_{-\infty}^{\infty} e^{-2\tau} u(\tau) 2e^{-(t-\tau)} u(t - \tau) d\tau \\ &= 2e^{-t} \int_{-\infty}^{\infty} e^{-\tau} u(\tau) u(t - \tau) d\tau = 2e^{-t} \left[\int_0^t e^{-\tau} d\tau \right] u(t) \\ &= 2(e^{-t} - e^{-2t}) u(t) \end{aligned}$$

(b) Using the above $h(t)$, we have

$$\begin{aligned}\int_{-\infty}^{\infty} |h(\tau)| d\tau &= 2 \int_0^{\infty} (e^{-\tau} - e^{-2\tau}) d\tau = 2 \left[\int_0^{\infty} e^{-\tau} d\tau - \int_0^{\infty} e^{-2\tau} d\tau \right] \\ &= 2 \left(1 - \frac{1}{2} \right) = 1 < \infty\end{aligned}$$

Thus, the system is BIBO stable.

Question 4

2.15. Consider a continuous-time LTI system with the input-output relation given by

$$y(t) = \int_{-\infty}^t e^{-(t-\tau)} x(\tau) d\tau \quad (2.82)$$

(a) Find the impulse response $h(t)$ of this system.

(b) Show that the complex exponential function e^{st} is an eigenfunction of the system.

(c) Find the eigenvalue of the system corresponding to e^{st} by using the impulse response $h(t)$ obtained in part (a).

SOLUTION

(a) From Eq. (2.82), definition (2.1), and Eq. (1.21) we get

$$\begin{aligned}h(t) &= \int_{-\infty}^t e^{-(t-\tau)} \delta(\tau) d\tau = e^{-(t-\tau)} \Big|_{\tau=0} = e^{-t} \quad t > 0 \\ h(t) &= e^{-t} u(t)\end{aligned} \quad (2.83)$$

Thus,

(b) Let $x(t) = e^{st}$. Then

$$\begin{aligned}y(t) &= \int_{-\infty}^t e^{-(t-\tau)} e^{s\tau} d\tau = e^{-t} \int_{-\infty}^t e^{(s+1)\tau} d\tau \\ &= \frac{1}{s+1} e^{st} = \lambda e^{st} \quad \text{if } \operatorname{Re} s > -1\end{aligned} \quad (2.84)$$

Thus, by definition (2.22) e^{st} is the eigenfunction of the system and the associated eigenvalue is